

الجامعة السورية الخاصة SYRIAN PRIVATE UNIVERSITY

كلية هندسة الحاسوب والمعلوماتية **Computer and Informatics Engineering** Faculty

Electric Circuits I

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Chapter 5 Operational Amplifier

- 5.1 What is an Op Amp?
- 5.2 Ideal Op Amp
- 5.3 Configuration of Op Amp
- 5.4 Cascaded Op Amp
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5.1 What is an Op Amp

- □ The **operational amplifier**, or (**op amp** for short) is an electronic unit that behaves like a voltage-controlled voltage source (VCVS).
- □ It is an active circuit element designed to perform mathematical operations of *addition*, *subtraction*, *multiplication*, *division*, *differentiation* and *integration*.
- □ Op amps are *commercially* available in *integrated circuit* (IC) *packages* in several forms as shown in Fig. for a typical operational amplifier.



Op Amp

- A typical Op amp is the eight-pin dual in-line package (or DIP), Fig. (a).
- Pin or terminal 8 is unused, and terminals 1 and 5 are of little concern to us.
- The **five important terminals** are:
 - The inverting input, pin 2.
 - The noninverting input, pin 3.
 - The output, pin 6.
 - The positive power supply V^+ , pin 7
 - The negative power supply V, pin 4.
- The **circuit symbol** for the op amp is the triangle in Fig.(b);
 - The inputs are marked with minus (-) and plus (+) to specify *inverting* and *noninverting* inputs.



Op Amp

- As an active element, the **op amp** must be powered by a voltage supply (V_{CC}) , Fig. (a)
 - By applying KCL:

 $\dot{i}_o = \dot{i}_1 + \dot{i}_2 + \dot{i}_+ + \dot{i}_-$

- The **equivalent circuit model** of an op amp, Fig.(b).
 - R_i is the Thevenin equivalent resistance seen at the input terminals, while R_o is the Thevenin equivalent resistance seen at the output.
 - The differential input voltage v_d is given by

where v_1 – inverting terminal voltage; v_2 – v_1 noninverting terminal voltage.

• The output v_o is given by $v_o = Av_d = A(v_2 - v_1)$



• When there is a feedback path from output to input, the ratio of the output voltage to the input voltage is called the **closed-loop gain**.

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Op Amp

Typical ranges for op amp parameters (see table).

- The op amp can operate in three modes, depending on the differential input voltage v_d , as shown in Fig.
 - 1. Positive saturation, $v_o = V_{CC}$.
 - 2. Linear region,
 - $-V_{\rm CC} \le v_o = A v_d \le V_{\rm CC}.$
 - 3. Negative saturation, $v_o = -V_{CC}$
- We will assume that our op amps operate in the linear mode, the output voltage is restricted by

 $-V_{CC} \le V_o \le V_{CC}$

Parameter	Typical range	Ideal values
Open-loop gain,	A 10^5 to $10^8 \Omega$	$\Omega \propto$
Input resistance,	$R_{\rm i}$ 10 ⁵ to 10 ¹³ Ω	$\Omega \propto$
Output resistance,	$R_{\rm o}$ 10 to 100 Ω	0 Ω
Supply voltage, V	V _{CC} 5 to 24 V	



Example 5.1

A 741 op amp has an open-loop voltage gain of 2×10^5 , input resistance of 2 M Ω , and output resistance of 50 Ω . The op amp is used in the circuit of Fig.(a).

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(a)

741

↓ i

10 kΩ

- a) Find the closed-loop gain  $v_0/v_s$ .
- b) Determine current *i* when  $v_s = 2$  V.

#### **Solution**

Using the equivalent circuit model for op amp, we obtain the equivalent circuit of Fig.(a) as shown in Fig.(b). At **node 1**, KCL gives

$$\frac{v_{s} - v_{1}}{10 \times 10^{3}} = \frac{v_{1}}{2000 \times 10^{3}} + \frac{v_{1} - v_{o}}{20 \times 10^{3}} \Rightarrow 200v_{s} = 301v_{1} - 100v_{o} \Rightarrow 2v_{s} \approx 3v_{1} - v_{o}$$

$$\Rightarrow v_{1} = \frac{2v_{s} + v_{o}}{3} (1)$$
At node  $O$ ,  $\frac{v_{1} - v_{o}}{20 \times 10^{3}} = \frac{v_{o} - Av_{d}}{50}$ , but  $v_{d} = v_{1}$   

$$\Rightarrow v_{1} - v_{o} = 400[v_{o} + (2 \times 10^{5})v_{1}] (2)$$
Substituting  $v_{1}$  from Eq. (1) into Eq. (2) gives  
 $0 \approx 26667067v_{o} + 5333333v_{s} \Rightarrow \frac{v_{o}}{v_{s}} = -1.9999699$ 
(b)  
 $v_{s} = 2V \Rightarrow v_{o} = -3.9999398V$  and  $v_{1} = 20.0666667V \Rightarrow i = \frac{v_{1} - v_{o}}{20 \times 10^{3}} = 0.19999 \text{ mA}$ 
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### 5.2 Ideal Op Amp

An ideal op amp has the following characteristics:

- Infinite open-loop gain,  $A \approx \infty$
- Infinite input resistance,  $R_i \approx \infty$
- Zero output resistance,  $R_0 \approx 0$ .

**Two important characteristics** of the ideal op amp are:
1. The currents into both input terminals are zero:

$$i_1 = 0, \quad i_2 = 0$$

2. The voltage across the input terminals is equal to zero; i.e.,

$$v_d = v_2 - v_1 = 0 \implies v_1 = v_2$$

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 $v_d$ 

 $v_2 = v_1$ 

### Example 5.2.

Using the ideal op amp model in Fig.

- a) Find the closed-loop gain  $v_0/v_s$ .
- b) Determine current  $i_o$  when  $v_s = 1$  V. *Solution:*
- Notice that  $v_2 = v_s$
- Since *i*<sub>1</sub> = 0, the 40-kΩ and 5-kΩ resistors are in series; The same current flows through them.
- $v_1$  is the voltage across the 5-k $\Omega$  resistor.
- Hence, using the voltage division principle (VDR),
   5 v

$$v_1 = \frac{5}{5+40} v_o = \frac{v_o}{9}$$

• For ideal op amp, we know that  $v_2 = v_1$ 

So, 
$$v_2 = v_1 = v_s = \frac{v_o}{9} \implies \frac{v_o}{v_s} = 9$$

• At node *O*, (KCL):

$$i_o = \frac{V_o}{40+5} + \frac{V_o}{20}$$
 mA

$$v_{2} \stackrel{i_{2} = 0}{\downarrow} \qquad \downarrow i_{1} = 0 \qquad \downarrow i_{0} \qquad \downarrow i_{0$$

• When  $v_s = 1$  V  $\rightarrow v_o = 9$  V. Then,

$$i_o = 0.2 + 0.45 = 0.65$$
 mA

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# 5.3 Configuration of Op amp

- 1. Inverting amplifier reverses the polarity of the input signal while amplifying it.
  - □ The circuit of inverting amplifier is shown in Fig.(a).
    - The noninverting input is grounded,
    - $v_i$  is connected to the inverting input through  $R_{1,i}$
    - The feedback resistor  $R_f$  is connected between the inverting input and output.
  - Our goal is to obtain the remaining voltage  $v_0$ . A contribution voltage  $v_i$  and the output voltage  $v_0$ . A contribution of the product of the produc • Our goal is to obtain the relationship between the

    - But  $v_1 = v_2 = 0$  for an ideal op amp, since the noninverting terminal is grounded. Hence,

$$\frac{v_i}{R_1} = -\frac{v_o}{R_f} \implies v_o = -\frac{R_f}{R_1}v_i$$

$$A_{v} = \frac{v_o}{v_i} = -\frac{R_f}{R_1}$$

(b)

0 A

0 V

(a)

 $\geq R_1$ 

 $R_1$ 

- The voltage gain is
- The equivalent circuit for the inverting amplifier is shown in Fig. (b)

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#### Example 5.3.

Refer to the op amp in Fig. If  $v_i = 0.5V$ , calculate:

- (a) the output voltage, v<sub>o</sub>;
  (b) the current in the 10kΩ resistor.
  Solution:
- a) Output voltage:

$$v_o = -\frac{R_f}{R_1}v_i = -\frac{25}{10}(0.5) = -1.25$$
 V

b) the current through the  $10k\Omega$  resistor is:

$$i = \frac{v_i - 0}{R_1} = \frac{0.5 - 0}{10 \times 10^3} = 50\,\mu\text{A}$$

25 kΩ

 $\sim$ 

10 kΩ

Vi

#### Example 5.4.

Determine  $v_0$  in the op amp circuit shown in Fig. *Solution:* 

Applying KCL at node a,

$$\frac{v_a - v_o}{40 \,\mathrm{k}\Omega} = \frac{6 - v_a}{20 \,\mathrm{k}\Omega} \Longrightarrow v_a - v_o = 12 - 2v_a$$
$$\Rightarrow v_o = 3v_a - 12$$

But  $v_a = v_b = 2$  V for an ideal op amp. Hence,

$$v_o = 3 \times 2 - 12 = -6$$
 V

Notice that if  $v_b = 0 = v_a$ , then  $v_o = -12$ .



#### 2. Non-inverting amplifier is designed to produce positive voltage gain.

The circuit of the op amp is the noninverting amplifier is shown in Fig. • Application of KCL at the inverting terminal gives

 $A_{v} = \frac{V_{o}}{1} = 1 + 1$ 

$$\dot{i}_1 = \dot{i}_2 \Longrightarrow \frac{0 - v_1}{R_1} = \frac{v_1 - v_o}{R_f}$$

• But 
$$v_1 = v_2 = v_i$$

 $\frac{-v_i}{R_1} = \frac{v_i - v_o}{R_f} \Longrightarrow \qquad v_o = \left( \frac{1 + v_o}{R_f} \right)$ 

The voltage gain is

which does not have a negative sign. Thus, the output has the same polarity as the input.

- In noninverting amplifier circuit, if feedback resistor  $R_f = 0$  (short circuit) or  $R_1 = \infty$  (open circuit) or both, the gain becomes 1.
  - Under these conditions, the circuit is called a voltage follower (تابع/متتبع الجهد) (or unity gain amplifier). Hence,  $V_o = V_i$
  - Such a circuit has a very high input impedance and is therefore useful as an intermediate-stage (or **buffer**) amplifier (مكبر صناد للتعذية المرتدة) to isolate one circuit from another, as in Fig.



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 $a = v_i$ 

### Example 5.5.

For the op amp shown in Fig., calculate the output voltage  $v_o$ . Solution:

**METHOD 1.** Using superposition, we let

$$v_o = v_{o1} + v_{o2}$$

where  $v_{01}$  is due to the 6-V voltage source, and  $v_{02}$  is due to the 4-V input.

• To get  $v_{01}$ , we set the 4-V source equal to zero. The circuit becomes an invert. amp.

$$=-\frac{10}{4}(6)=-15$$
 V

6 \

- To get  $v_{02}$ , we set the 6-V source equal to zero. The circuit becomes a  $v_{02} = \left(1 + \frac{10}{\Lambda}\right) 4 = 14 \text{ V}$ noninvert. amp., so that
- Thus,  $v_o = v_{o1} + v_{o2} = -15 + 14 = -1$  V

**METHOD 2.** Using nodal analysis.

• Applying KCL at node *a*,

$$\frac{6 - v_a}{4} = \frac{v_a - v_o}{10}$$

But  $v_a = v_b = 4$ , so  $5 = 4 - v_o \implies v_o = -1$  V

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10 kΩ

~~~~

a

4 V

- **3. Summing Amplifier** is an op amp circuit that combines several inputs and produces an output that is the weighted sum of the inputs.
 - Applying KCL at node *a* gives:

 $i = i_1 + i_2 + i_3$

• But $i_1 = \frac{v_1 - v_a}{R_1}; i_2 = \frac{v_2 - v_a}{R_2}$

$$i_3 = \frac{v_3 - v_a}{R_3}; \ i = \frac{v_a - v_o}{R_f}$$

• We note that $v_a = 0$, thus

$$v_o = -\left(\frac{R_f}{R_1}v_1 + \frac{R_f}{R_2}v_2 + \frac{R_f}{R_3}v_3\right)$$

a

Example 5.6.

Calculate v_o and i_o in the op amp circuit shown below.

Solution:

This is a summer with two inputs.

$$v_o = -\left[\frac{10}{5}(2) + \frac{10}{2.5}(1)\right] = -(4+4) = -8V$$

- The current i_o is the sum of the currents through the 10-k Ω and 2-k Ω resistors.
- Both of these resistors have voltage $v_o = -8$ V across them, since $v_a = v_b = 0$.
 - Hence, $i_0 = \frac{v_o 0}{10} + \frac{v_o 0}{2} = -0.8 4 = -4.8 \text{ mA}$



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10 kΩ

10

2 kΩ

а

b

 $5 k\Omega$

2 V

2.5 kΩ

1 V

4. **Difference amplifier** is a device that amplifies the difference between two inputs but rejects any signals common to the two inputs.



$$\frac{v_1 - v_a}{R_1} = \frac{v_a - v_o}{R_2} \Longrightarrow v_o = \left(\frac{R_2}{R_1} + 1\right) v_a - \frac{R_2}{R_1} v_1$$

• Applying KCL to node *b*,

$$\frac{v_2 - v_b}{R_3} = \frac{v_b - 0}{R_4} \Longrightarrow v_b = \frac{R_4}{R_3 + R_4} v_2$$

• But $v_a = v_b$, thus

or
$$v_o = \left(\frac{R_2}{R_1} + 1\right) \frac{R_4}{R_3 + R_4} v_2 - \frac{R_2}{R_1} v_3$$

vь

R₄

 R_{2}

$$v_o = \frac{R_2 (1 + R_1 / R_2)}{R_1 (1 + R_3 / R_4)} v_2 - \frac{R_2}{R_1} v_1$$

 R_{2} R_{2}

if
$$\frac{R_2}{R_1} = \frac{R_3}{R_4} = 1 \implies v_o \equiv v_2 - v_2$$

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Example 5.7.

Determine R_1 , R_2 , R_3 and R_4 so that $v_0 = -5v_1 + 3v_2$ for the circuit shown below.

 R_2

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Solution:

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The output for this amplifier is $v_o = \frac{R_2(1+R_1/R_2)}{R_1(1+R_2/R_1)}v_2 - \frac{R_2}{R_1}v_1$ v_{h} v_1 R_4 Comparing with $v_{o} = 3v_{2} - 5v_{1}$ $\frac{R_2}{R_1} = 5 \Longrightarrow R_2 = 5R_1$ we see that Also, $5\frac{(1+R_1/R_2)}{(1+R_3/R_4)} = 3 \implies (1+R_3/R_4) = \frac{3}{5}$ $2 = 1 + \frac{R_3}{R_4} \implies R_3 = R_4$ Or If we choose $R_1 = 10$ k Ω and $R_3 = 20$ k Ω , then $R_2 = 50$ k Ω and $R_4 = 20$ k Ω .

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5.4 Cascaded Op Amp

- A cascade connection is a head-to-tail arrangement of two or more op amp circuits such that the output to one is the input of the next.
 - When op amp circuits are cascaded, each circuit in the string is called a *stage*.
 - The original input signal is *increased* by the *gain* of the individual stage.
 - Figure displays a block diagram representation of three op amp circuits in cascade.
 - Since the output of one stage is the input to the next stage, the overall gain of the cascade connection is the product of the gains of the individual op amp circuits, or $A = A_1 A_2 A_3$

+ Stage 1 +
$$v_1$$
 A₁ + $v_2 = A_1v_1$ A₂ + $v_3 = A_2v_2$ A₃ + $v_o = A_3v_3$

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Example 5.8.

Find v_0 and i_0 in the circuit shown in Fig. **Solution:**

This circuit consists of two noninverting amplifiers cascaded.

• At the output of the first op amp,

$$v_a = \left(1 + \frac{12}{3}\right)(20) = 100 \text{ mV}$$

- At the output of the second op amp, $v_a = \left(1 + \frac{10}{4}\right)v_a = 350 \text{ mV}$
- The required current i_o is the current through the 10-k Ω resistor: $i_o = \frac{v_o v_b}{10}$ mA
- But $v_b = v_a = 100 \text{ mV}$

• Hence,
$$i_o = \frac{(350 - 100) \times 10^{-3}}{10 \times 10^3} = 25 \mu \text{A}$$

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a

12 kΩ

3 kΩ

20 mV

b

10 kΩ

4 kΩ

vo

Example 5.9.

Find v_0 in the op amp circuit of Fig.

Solution:

- Let v_1 output of the first op amp and v_2 output of the second op amp.
- The *first stage* is an **inverting** amplifier:
- The *second state* is a summer:

$$v_2 = -\left(\frac{100}{50}v_{s2} + \frac{100}{100}v_1\right) = -2v_{s2} - v_1 = -2v_{s2} + 2v_{s1}$$

• The *third state* is a **noninverting** amplifier:

$$v_o = \left(1 + \frac{100}{50}\right)v_2 = 3v_2 = 3(-2v_{s2} + 2v_{s1}) = 6v_{s1} - 6v_{s2}$$

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25 kΩ

~~~~

 $v_1 =$ 

Vs1 O-

50 kΩ

 $\frac{50}{25}v_{s1}$ 

50 kΩ

100 kΩ

 $100 \text{ k}\Omega$ 

100 kΩ

50 kΩ

## 5.5 Application: Digital-to Analog Converter

- Thedigital-to-analogconverter(DAC)transforms digital signals intoanalog form.
  - A typical example of a four-bit DAC is illustrated in Fig.(a).
  - The **four-bit DAC** can be realized in many ways.
  - A simple realization is the binary weighted ladder ( شلم = درج موزون), Fig. (b).

$$-V_0 = \frac{R_f}{R_1}V_1 + \frac{R_f}{R_2}V_2 + \frac{R_f}{R_3}V_3 + \frac{R_f}{R_4}V_4$$



where:  $V_1 - MSB$  (most significant bit),  $V_4 - LSB$  (least significant bit );

 $V_1$  to  $V_4$  are either 0 or 1 V.

#### Example 5.10.

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In the op amp circuit of Fig. , let  $R_f = 10 \text{ k}\Omega$ ,  $R_1 = 10 \text{ k}\Omega$ ,  $R_2 = 20 \text{ k}\Omega$ ,  $R_3 = 40 \text{ k}\Omega$ , and  $R_4 = 80 \text{ k}\Omega$ . obtain the analog output for binary inputs [0000], [0001], [0010], ..., [1111].

Substituting the given values of the input and feedback resistors in following Eq.

$$V_{0} = \frac{R_{f}}{R_{1}}V_{1} + \frac{R_{f}}{R_{2}}V_{2} + \frac{R_{f}}{R_{3}}V_{3} + \frac{R_{f}}{R_{4}}V_{4} = V_{1} + 0.5V_{2} + 0.25V_{3} + 0.125V_{4}$$

Using this equation, a digital inputs produce an analog output as following:

$$\begin{bmatrix} V_1 V_2 V_3 V_4 \end{bmatrix} = \begin{bmatrix} 0000 \end{bmatrix} \Rightarrow -V_o = 0 \text{ V}$$
  

$$\begin{bmatrix} V_1 V_2 V_3 V_4 \end{bmatrix} = \begin{bmatrix} 0001 \end{bmatrix} \Rightarrow -V_o = 0.125 \text{ V}$$
  

$$\begin{bmatrix} V_1 V_2 V_3 V_4 \end{bmatrix} = \begin{bmatrix} 0010 \end{bmatrix} \Rightarrow -V_o = 0.25 \text{ V}$$
  

$$\begin{bmatrix} V_1 V_2 V_3 V_4 \end{bmatrix} = \begin{bmatrix} 0011 \end{bmatrix} \Rightarrow -V_o = 0.25 + 1.125 = 0.375 \text{ V}$$
  

$$\begin{bmatrix} V_1 V_2 V_3 V_4 \end{bmatrix} = \begin{bmatrix} 0100 \end{bmatrix} \Rightarrow -V_o = 0.5 \text{ V}$$

 $[V_1V_2V_3V_4] = [1111] \Rightarrow -V_o = 1 + 0.5 + 0.25 + 0.125 = 1.875 \text{ V}$ 

 $\leq R_2 \leq R_3$ 

LSB

MSB

Input and output values of the four-bit DAC.

| Binary input        |               | Output   |  |
|---------------------|---------------|----------|--|
| $[V_1 V_2 V_3 V_4]$ | Decimal value | $-V_{a}$ |  |
| 0000                | 0             | 0        |  |
| 0001                | 1             | 0.125    |  |
| 0010                | 2             | 0.25     |  |
| 0011                | 3             | 0.375    |  |
| 0100                | 4             | 0.5      |  |
| 0101                | 5             | 0.625    |  |
| 0110                | 0             | 0.75     |  |
| 0111                |               | 0.875    |  |
| 1000                | 8             | 1.0      |  |
| 1001                | 9             | 1.125    |  |
| 1010                | 10            | 1.25     |  |
| 1011                | 11            | 1.375    |  |
| 1100                | 12            | 1.5      |  |
| 1101                | 13            | 1.625    |  |
| 1110                | 14            | 1.75     |  |
| 1111                | 15            | 1.875    |  |
|                     |               |          |  |



# The end of chapter 5